

Lecture 11b

Metastability

Lena Peterson
2017-10-03

Why care?

- Digital abstraction:
 - All signals in a system must have a valid digital representation
- Consequence: Must reliably synchronize external events

Real world problem!

- Inputs from real world are usually not synchronous with system clock.
- Inputs from other synchronous systems are based on another system clock usually not synchronous with your system clock.

The metastability equation

$$MTBF = \frac{e^{\frac{S}{\tau}}}{T_W f_c f_D}$$

MTBF is the figure of merit for synchronisation

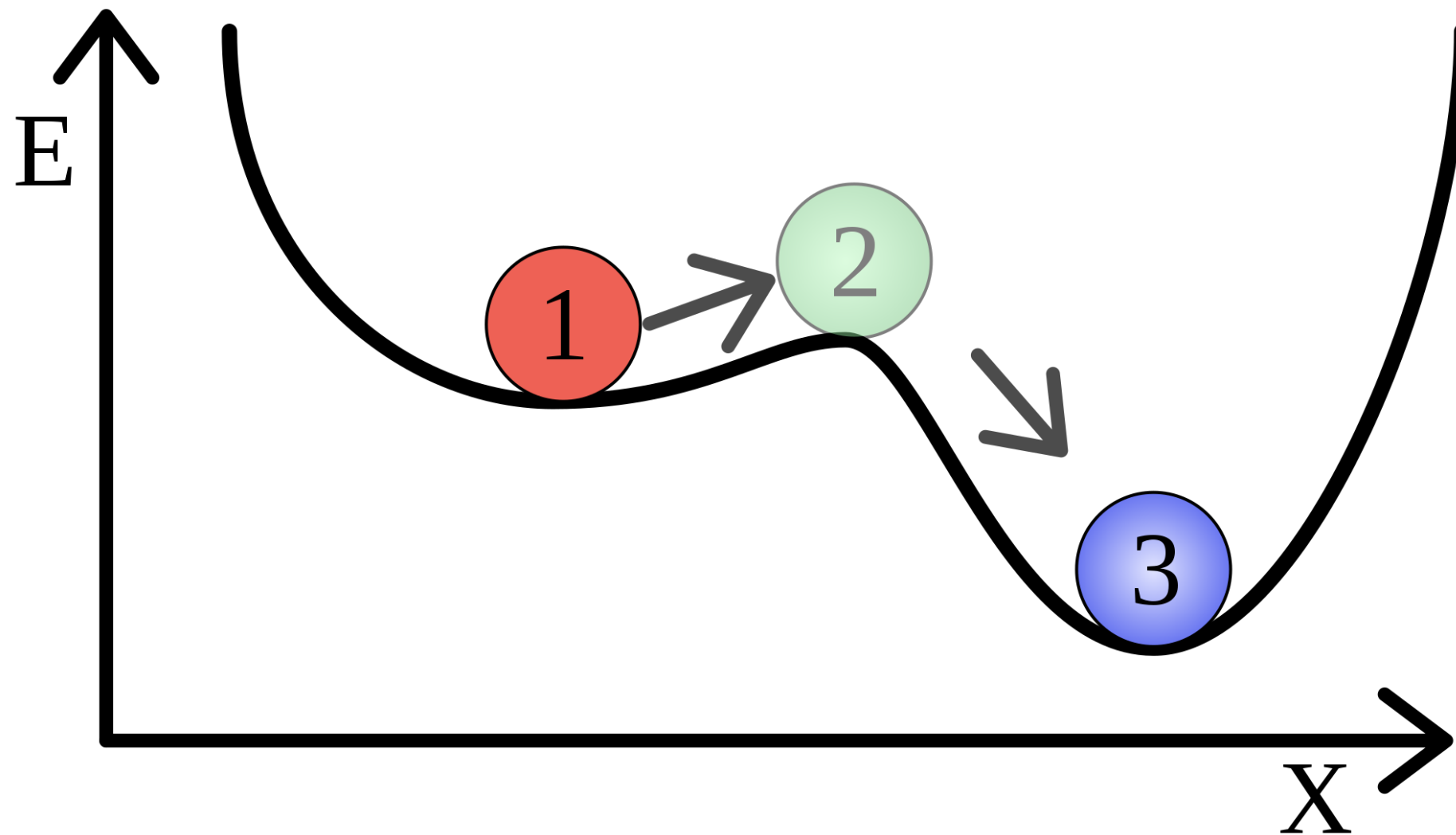
In this lecture we will derive this equation

See how problem can be alleviated

Derivation follows: “Metastability and Synchronizers: A tutorial” by R. Ginosar 2011

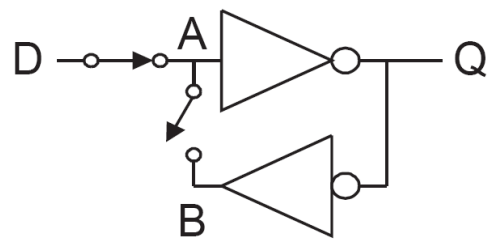
Paper is available in PingPong

Metastability

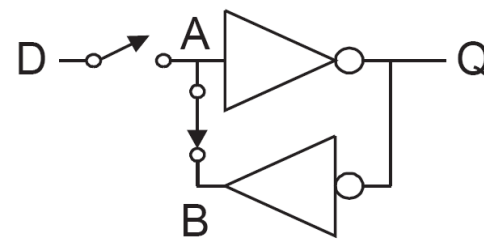


meta means “middle” or “in between”

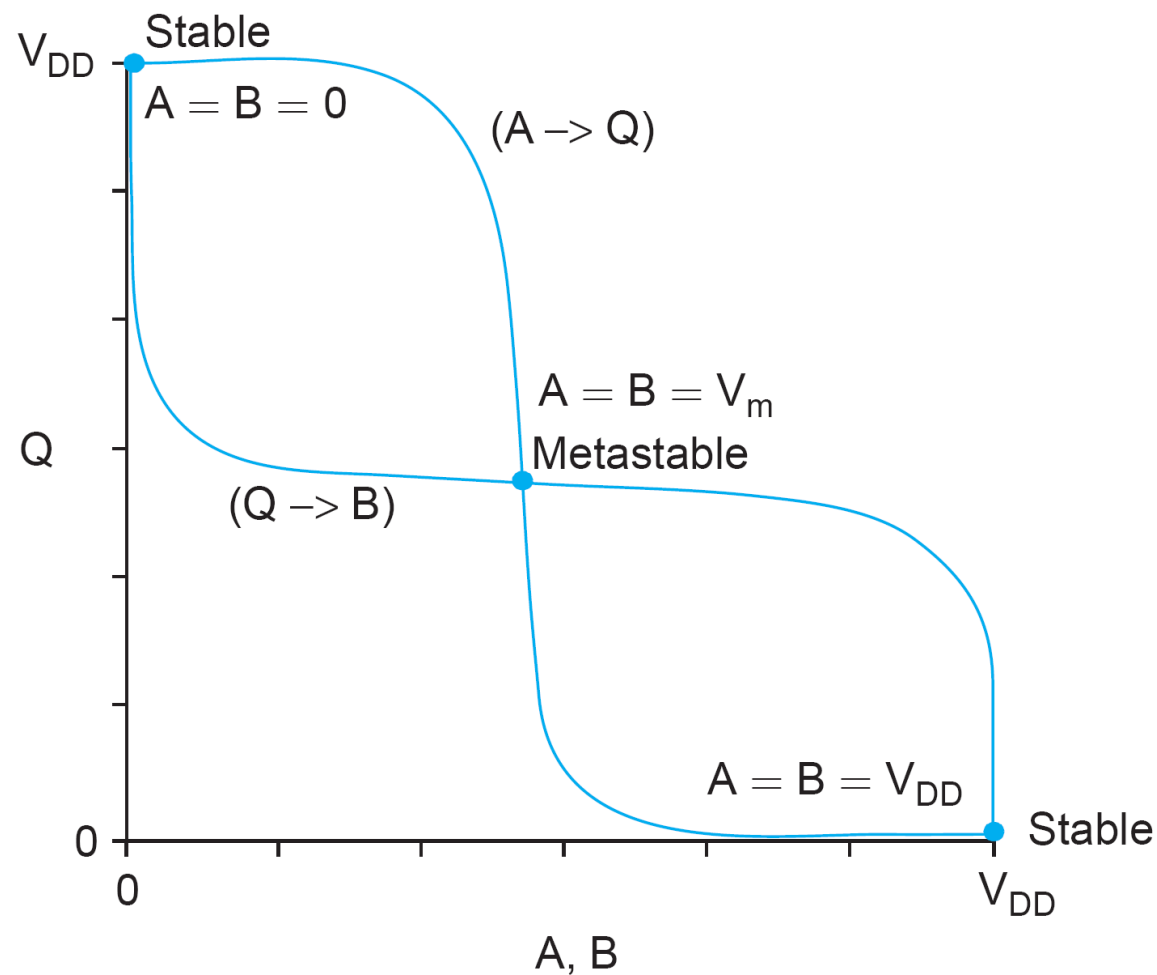
Metastability in latch



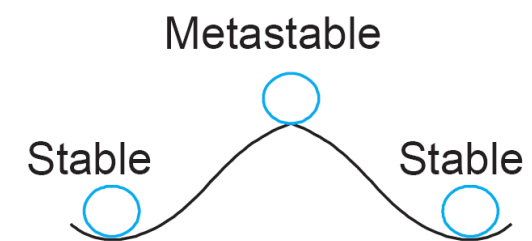
(a)



(b)



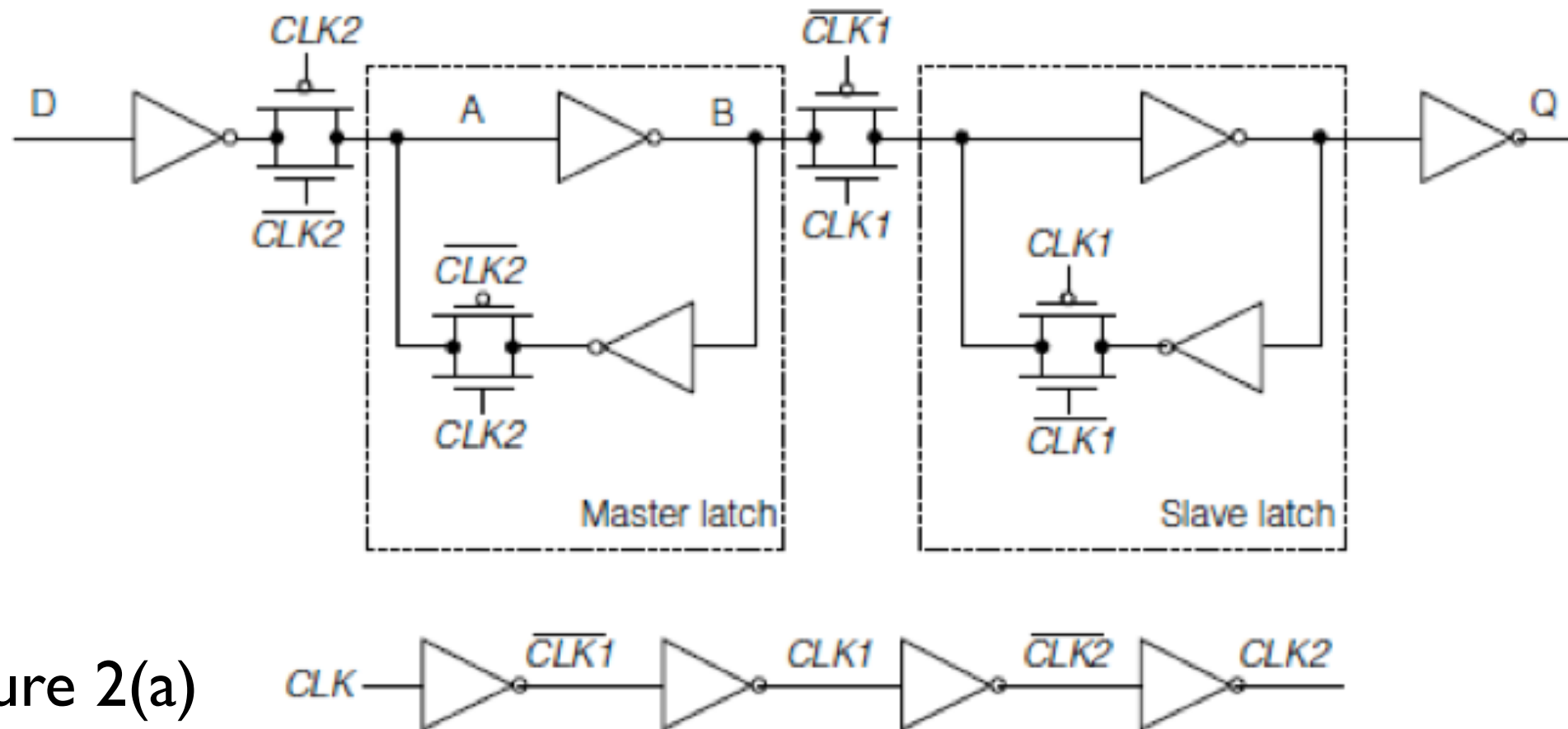
(c)



(d)

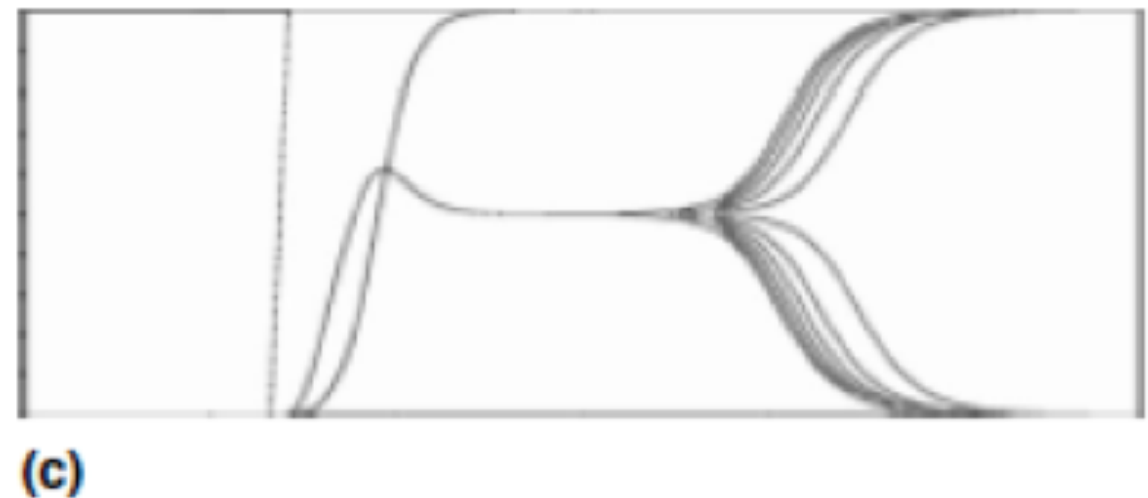
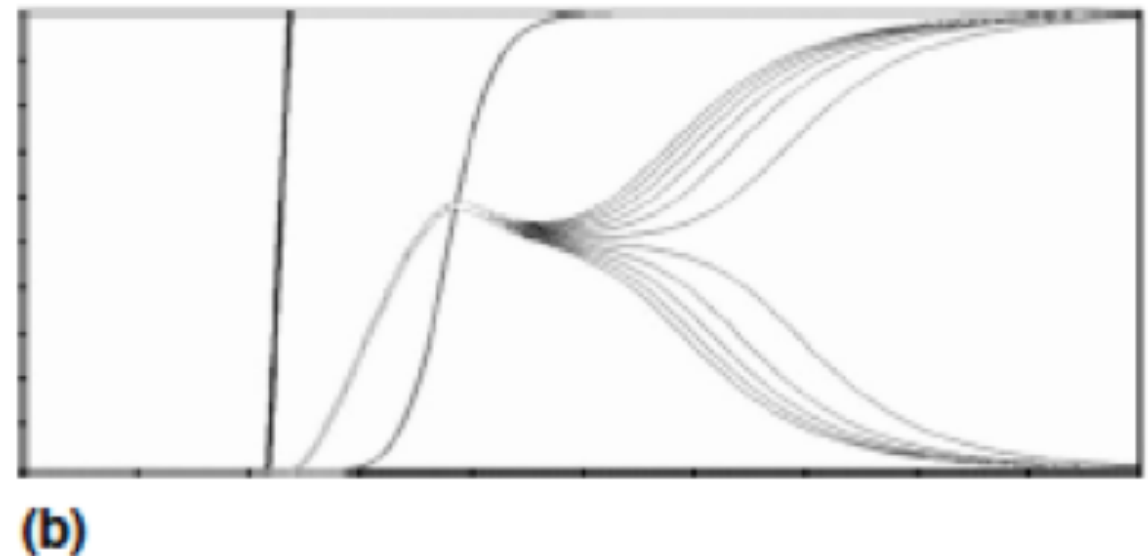
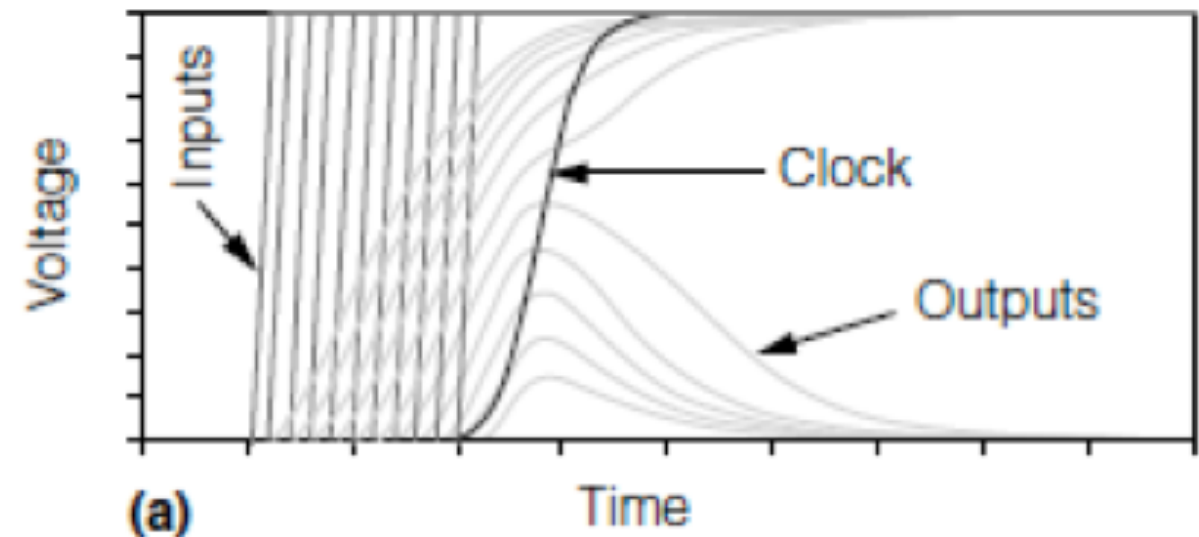
Figure:

Flip-flop = two latches



Entering metastability

Figure 3. Empirical circuit simulations of entering metastability in the master latch of Figure 2 (a). Charts show multiple inputs D, internal clock (CLK2) and multiple corresponding outputs Q. The input edge is moved in steps of (a) 100 ps, (b) 1 ps, and (c) 0.1 fs.



Entering metastability

$$Rate = f_D T_W f_c$$

f_c = clock frequency

T_w = window of metastability

f_D = data rate at flip-flop input

Another way to express it: $Rate = f_D \frac{T_W}{T_c}$

An example

- $f_c = 1 \text{ GHz}$
- $T_W = 20 \text{ ps}$
- $f_D = 100 \text{ MHz}$
- Rate for entering?

Exiting metastability

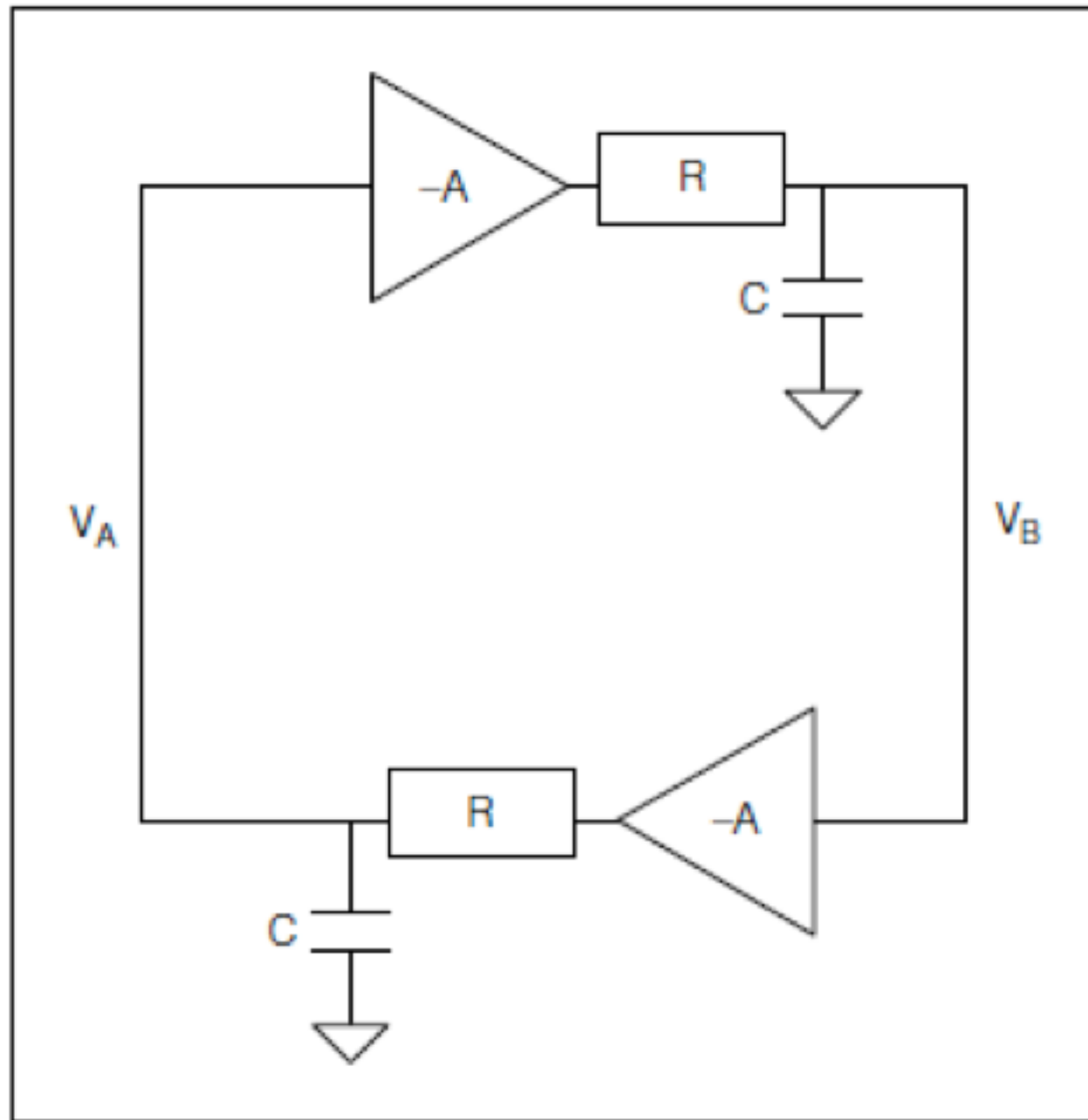
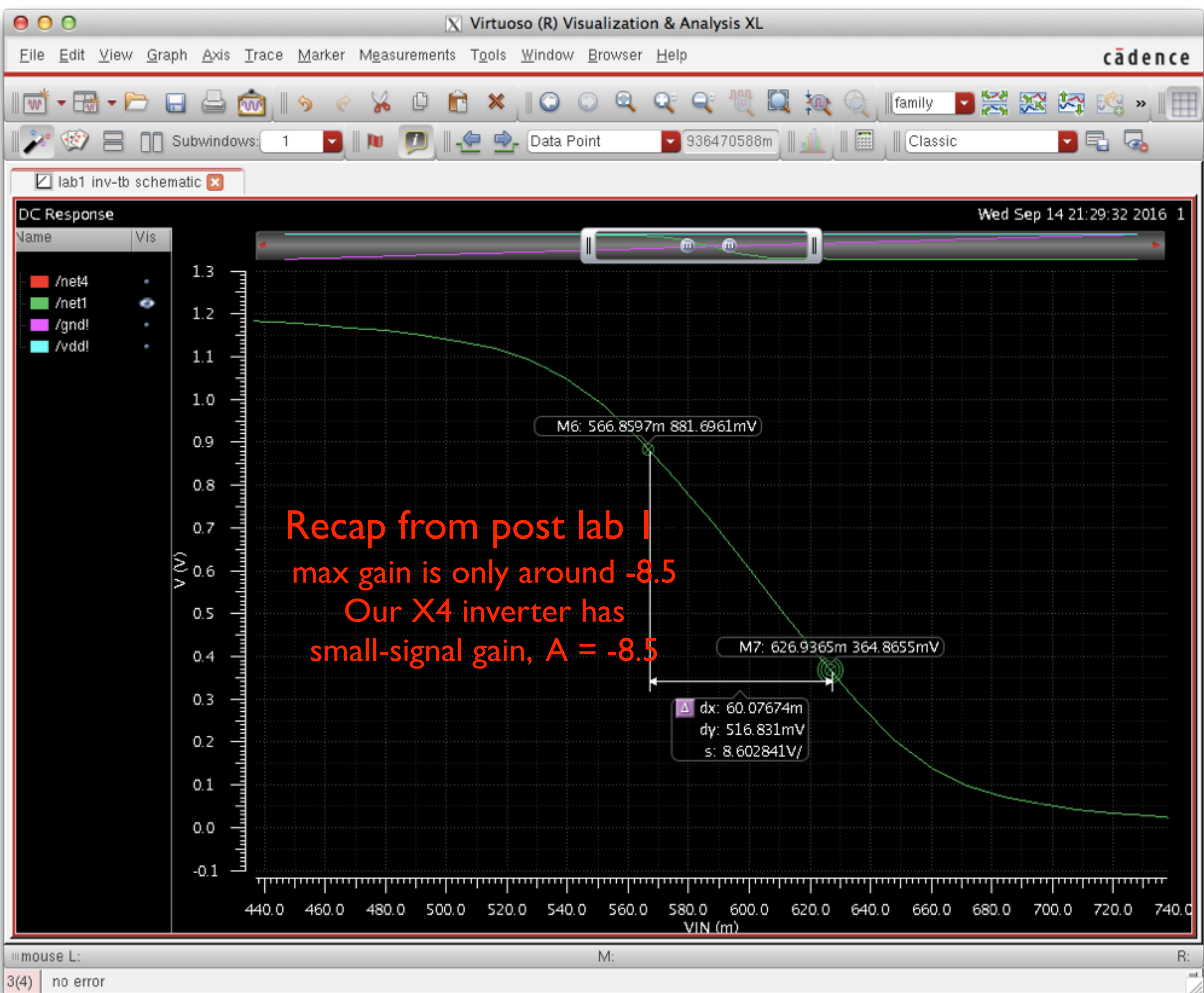


Figure 4. Analog model of a metastable latch; the inverters are modeled as negative amplifiers.

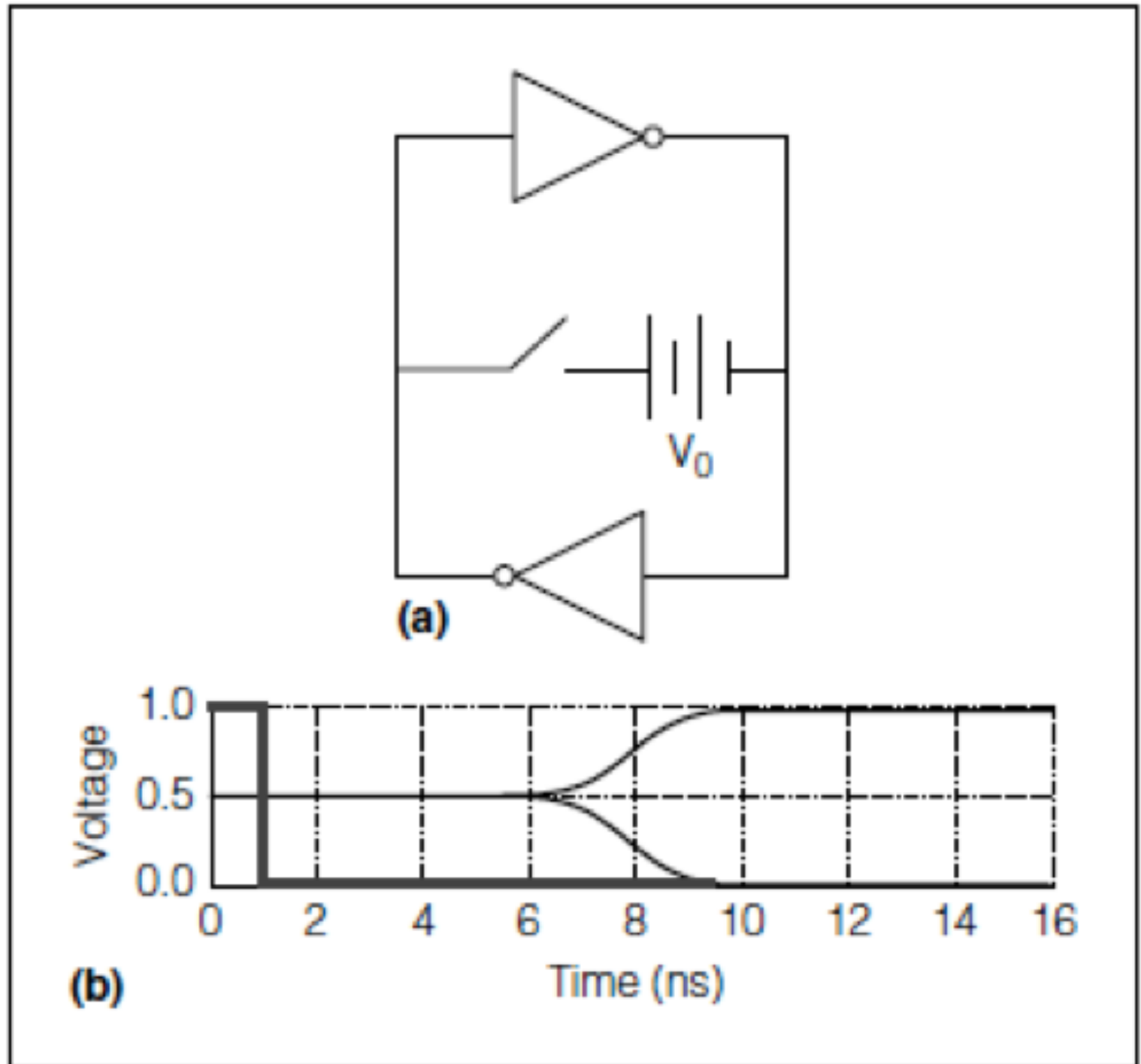
Small-signal model:,
 $-A$ is small-signal gain,
 R is output resistance
and C is input capacitance of other inverter.



Exiting

Figure 5. Simulation of exiting metastability: circuit (a) and voltage chart of the two latch nodes vs time (b).

The switch starts closed (applying $V_0 = 1\ \mu\text{V}$) and then opens up (at $t=1\ \text{ns}$) to allow latch to resolve.



Exiting metastability

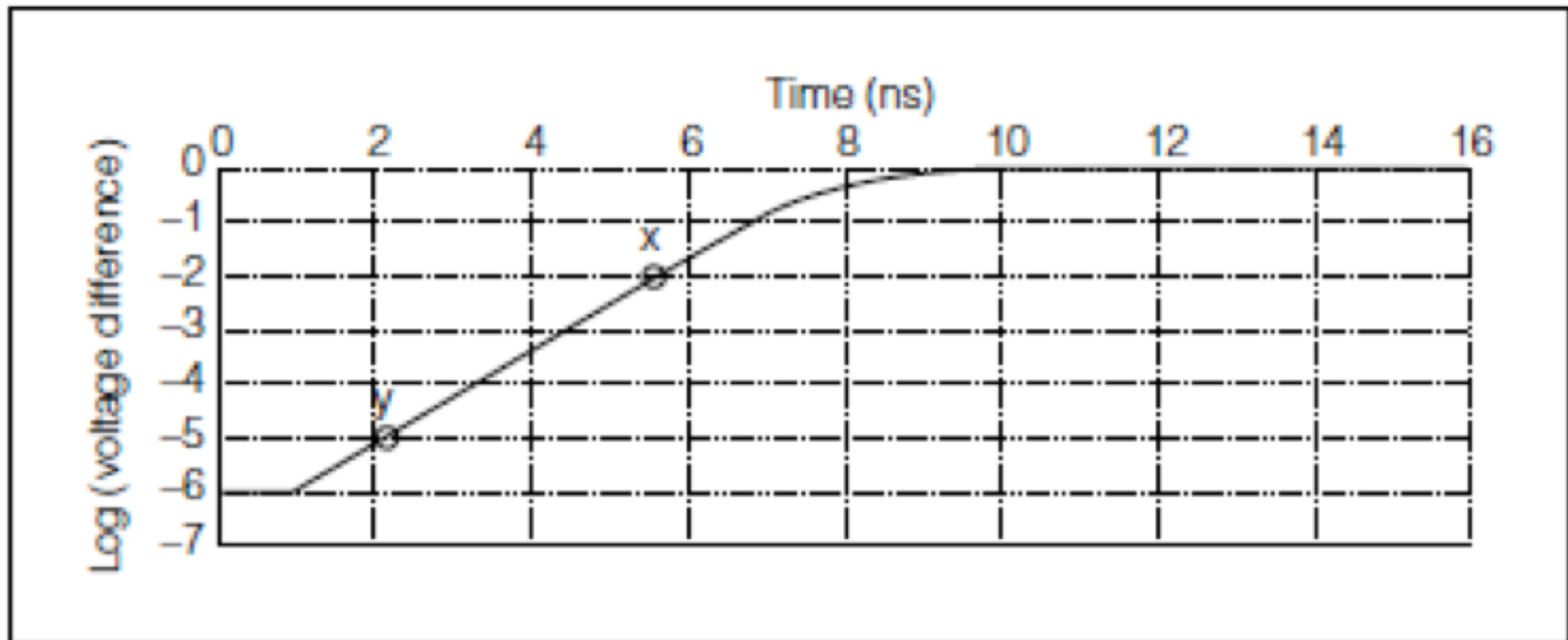
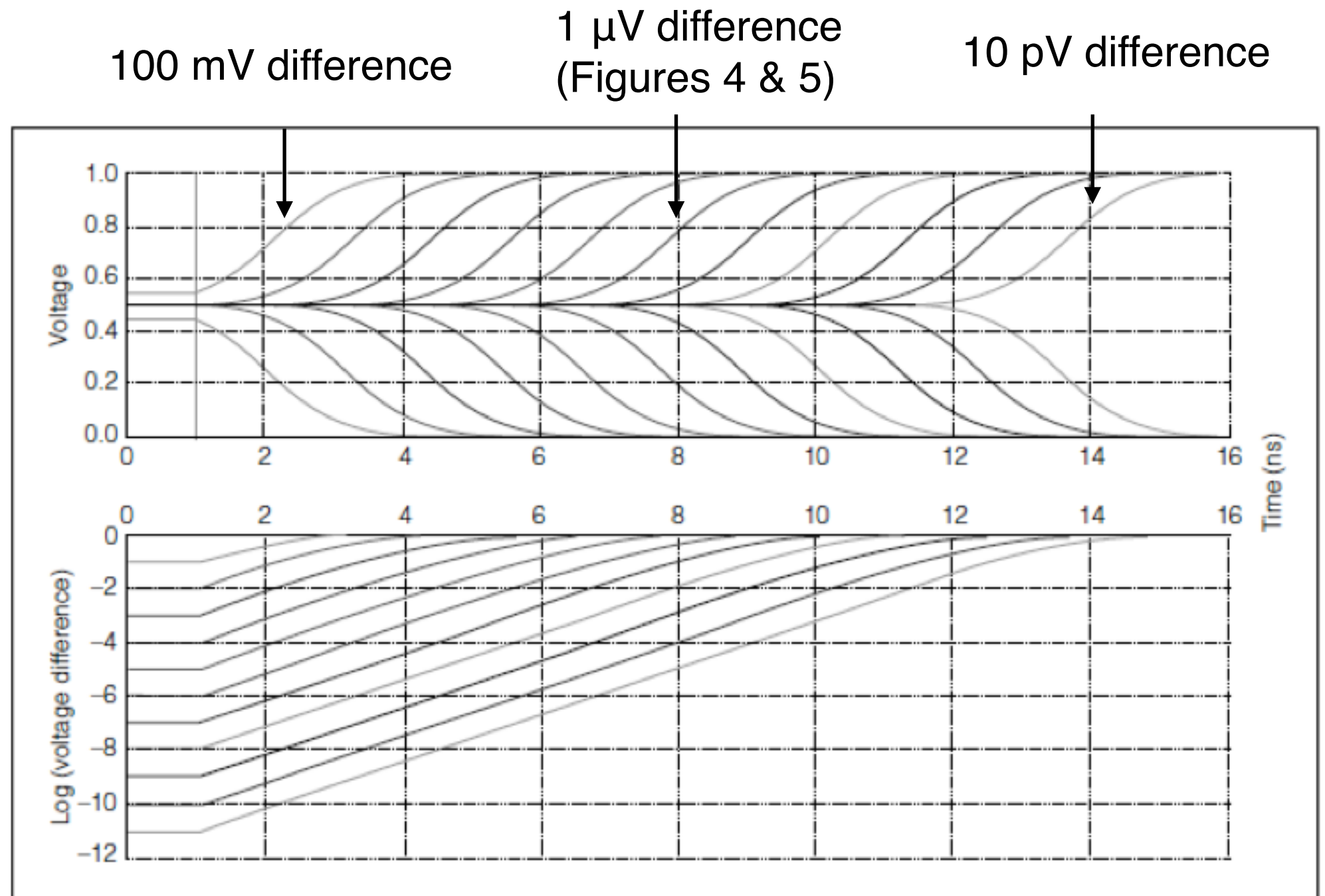


Figure 6. Log of the voltage difference of the two nodes of a resolving latch in Figure 5.

Points x and y can be used to determine time constant τ (assuming a straight line).

Exiting metastability

Figure 7.
Simulations of metastability resolution with the starting voltage difference varying from 100 mV (left) to 10 pV (right) in steps of factor 10; the lower the starting voltage difference the longer resolution takes.



Time to exit

$$V_1 = V_0 e^{\frac{t_m}{\tau}}$$

$$t_m = \tau \ln \left(\frac{V_1}{V_0} \right)$$

V_0 starting voltage, V_1 ending voltage,

t_m is time to exit metastability

τ depends on many things!

Often around 0.5 - 2 FO4 delays

Failure of synchronization

S is the allotted synchronization period

Probability of failure:

$$p(\text{failure}) = p(\text{entering}) \times p(\text{time to exit} > S)$$

$$Rate(failures) = T_W f_D f_c e^{-\frac{S}{\tau}}$$

$$MTBF = \frac{e^{\frac{S}{\tau}}}{T_W f_c f_D}$$

Avoiding synchronization failure

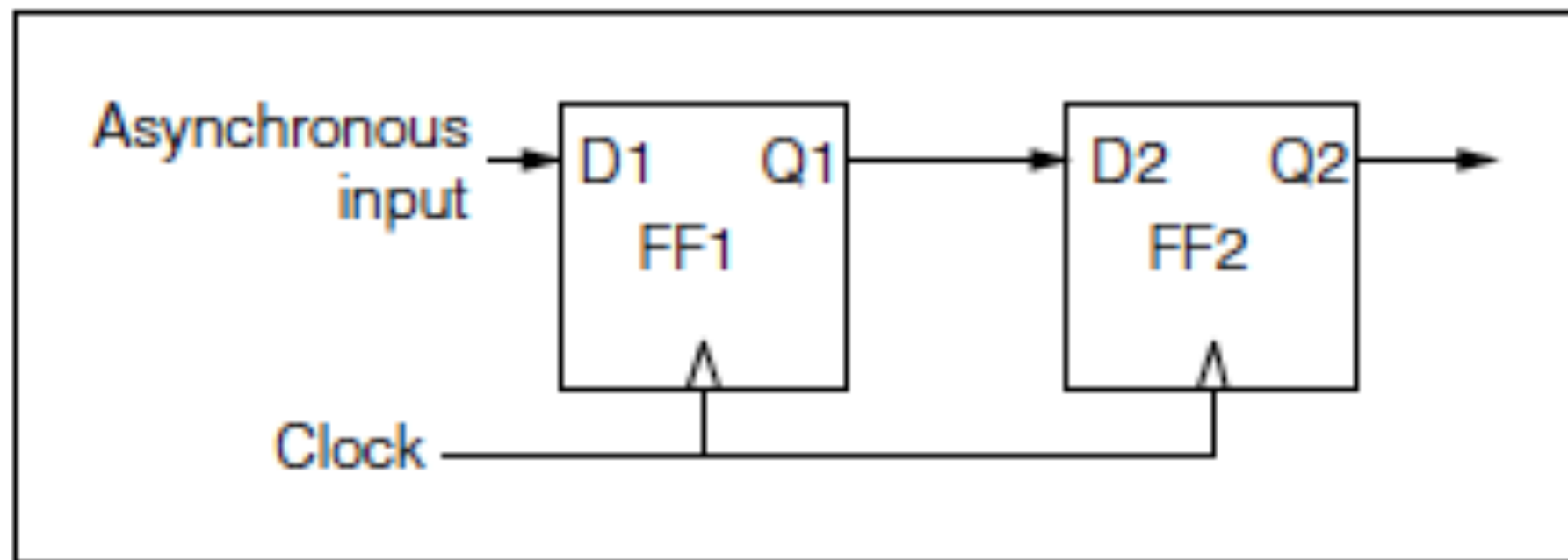


Figure 8. Two-flip-flop synchronization circuit.

6 cases

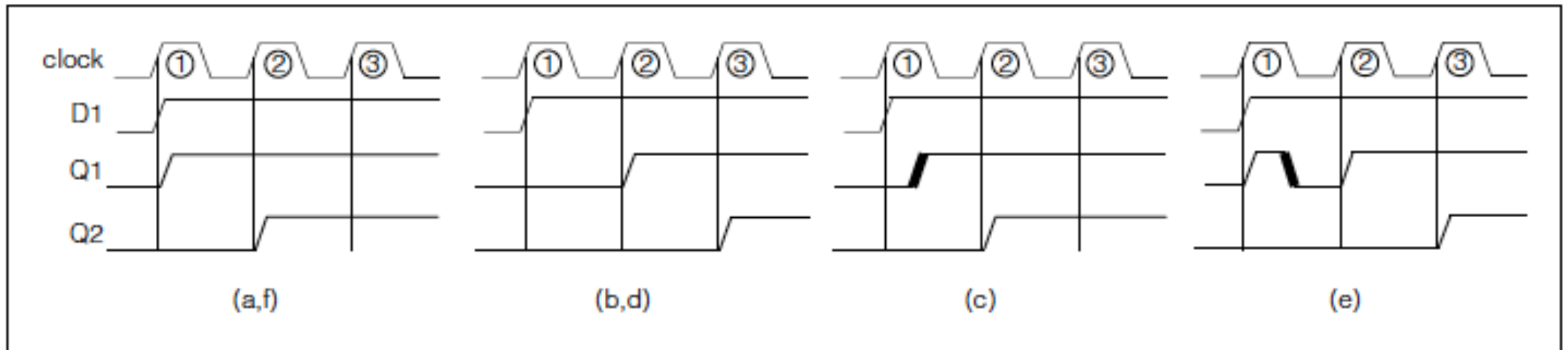


Figure 9. Alternative two-flip-flop synchronization waveforms.

Which are the 6 cases that can happen?

Many pitfalls

- Synchronizing the same signal more than once.
- Synchronizing buses
- Automated synchronizing solutions that do not work.
 - For example with dynamic latches

Metastability and synchronisation

- MTBF is figure of merit:
 - Probability to enter
 - Probability to exit
- Two FF synchronisers lower the MTBF
- Many pitfalls - be paranoid!